

## Lecture: 25 - 26

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### Topics: Basics of coding theory

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### Introduction

One of the major applications of finite fields is coding theory. This theory has its origin in a famous theorem of Shannon that guarantees the existence of code that can transmit information at rates close to capacity with an arbitrarily small probability of error. On purpose of algebraic coding theory, the theory of error-correcting and error-detecting codes, is to devise methods for the construction of such codes.

During the last two decades more and more abstract algebraic tools such as the theory of finite fields and polynomials over finite fields have influenced coding.

In particular the description on redundant codes by polynomials over  $\mathbb{F}_q$  is a milestone in this development. The fact that one can use shift registers for coding and decoding establishes a connection with linear recurring sequences.

### Linear Codes

The problem of the communication of information, in particular the coding and decoding of information for the reliable transmission over a "noisy" channel, is of great importance today.

Typically, one has to transmit a message which consists of a finite sequence of symbols that are elements of some finite alphabet. For instance, if this alphabet consists simply of 0 and 1, the message can be described as a binary number.

Generally the alphabet is assumed to be a finite field.

Now the transmission of finite sequences of elements of the alphabet over a communication channel need not be perfect in the sense that each bit of information is transmitted unaltered over this channel.

As there is no ideal channel without "noise" the receiver of the transmitted may obtain distorted information and may make errors in interpreting the transmitted signal.

One of the main problems of coding theory is to make the errors, which occur for instance because of noisy channels, extremely improbable. The methods to improve the reliability of transmission depend on properties of finite fields.

A basic idea in algebraic coding theory is to transmit redundant information together with the message one wants to communicate; that is one extends the sequence of message symbols to a longer sequence in a systematic manner.

A simple model of communication system is shown in figure.

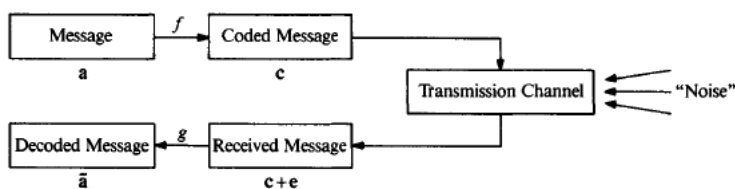


FIGURE 9.1

We assume that the symbols of the message and of the coded message are elements of the same finite field  $\mathbb{F}_q$ .

**Coding** means to encode a block of  $k$  message symbols  $a_1 a_2 \dots a_k$ ,  $a_i \in \mathbb{F}_q$  into a **code word**  $c_1 c_2 \dots c_n$  of  $n$  symbols  $c_j \in \mathbb{F}_q$  where  $n > k$ .

We regard the code word as an  $n$ -dimensional row vector  $\vec{c} \in \mathbb{F}_q^n$ .

Thus  $f$  in the figure is a function from  $\mathbb{F}_q^k$  into  $\mathbb{F}_q^n$  called a **coding scheme**, and  $g : \mathbb{F}_q^n \rightarrow \mathbb{F}_q^k$  is a **decoding scheme**.

A simple type of coding scheme arises when each block  $a_1 a_2 \dots a_k$  of message symbols is encoded into a code word of the form:

$$a_1 a_2 \dots a_k c_{k+1} \dots c_n$$

where the first  $k$  symbols are the original message symbols and the additional  $n - k$  symbols in  $\mathbb{F}_q$  are control symbols. Such coding schemes are often presented in the following way.

Let  $H$  be a given  $(n - k) \times n$  matrix with entries in  $\mathbb{F}_q$  that is of the special form:

$$H = (A, I_{n-k})$$

where  $A$  is an  $(n - k) \times k$  matrix and  $I_{n-k}$  the identity matrix of order  $n - k$ . The control symbols  $c_{k+1}, \dots, c_n$  can be calculated from the system of equations:

$$H\vec{c}^T = \vec{0}$$

for code words  $\vec{c}$ . The equations of this system are called **parity-check-equations**.

**Example**

Let  $H$  be the following  $3 \times 7$  matrix over  $\mathbb{F}_2$ :

$$H = [1 \ 0 \ 1 \ 1 \ 1 \ 0 \ 0 \ 1 \ 1 \ 0 \ 1 \ 0 \ 1 \ 0 \ 1 \ 1 \ 0 \ 0 \ 0 \ 1]$$

Then the control symbols can be calculated by solving:

$$H\vec{c}^T = \vec{0}$$

given  $c_1, c_2, c_3, c_4$ :

$$[c_1 \quad +c_3 \quad +c_4 \quad +c_5 \quad = 0 \quad c_1 \quad +c_2 \quad +c_4 \quad +c_6 \quad = 0 \quad c_1 \quad +c_2 \quad +c_3 \quad +c_7 \quad = 0]$$

The control symbols  $c_5, c_6, c_7$  can be expressed as:

$$[c_5 = c_1 + c_3 + c_4 \quad c_6 = c_1 + c_2 + c_4 \quad c_7 = c_1 + c_2 + c_3]$$

Thus the coding scheme in this case is a linear map from  $\mathbb{F}_2^4$  into  $\mathbb{F}_2^7$  given by:

$$(a_1, a_2, a_3, a_4) \rightarrow (a_1, a_2, a_3, a_4, a_1 + a_3 + a_4, a_1 + a_2 + a_4, a_1 + a_2 + a_3)$$

In general, we use the following terminology in connections with coding schemes that are given by linear maps.

**Definition 9.2**

Let  $H$  be an  $(n - k) \times n$  matrix of rank  $n - k$  with entries in  $\mathbb{F}_q$ .

The set  $C$  of all  $n$ -dimensional vectors  $\vec{c} \in \mathbb{F}_q^n$  such that  $H\vec{c}^T = \vec{0}$  is called a linear  $(n, k)$  code over  $\mathbb{F}_q$ .  $n$  is called the length  $k$  is called the dimension of the code.

The elements of  $C$  are called **code words** (or code vectors), the matrix  $H$  is a parity-check matrix of  $C$ .

If  $q = 2, C$  is called a binary code. If  $H$  is of the form  $(A, I_{n-k})$ , then  $C$  is called a systematic code.

Systematic code (from wiki): "In coding theory, a systematic code is any error-correcting code in which the input data is embedded in the encoded output. Conversely, in a non-systematic code the output does not contain the input symbols."

We note that the set  $C$  of solutions of the system  $H\vec{c}^T = \vec{0}$  of linear equations is a subspace of dimension  $k$  of the vector space  $\mathbb{F}_q^n$ .

Since the code words form an additive group,  $C$  is also called a **group code**. Moreover  $C$  can be regarded as the **null space** of the matrix  $H$ .

**Example 9.3 (Parity-check Code)**

Let  $q = 2$  and let the given message be  $a_1, a_2, \dots, a_k$  then the coding scheme  $f$  is defined by:

$$f : a_1 \dots a_k \rightarrow b_1 \dots b_{k+1}$$

where  $b_i = a_i$  for  $i = 1, \dots, k$  and:

$$b_{k+1} = \begin{cases} 0, & \text{if } \sum_{i=1}^k a_i = 0, \\ 1, & \text{if } \sum_{i=1}^k a_i = 1 \end{cases}$$

Hence it follows that the sum of digits of any code word  $b_1 \dots b_{k+1}$  is 0. If the sum of digits of the received word is 1. then the receiver knows that a transmission error must have occurred.

Let  $n = k + 1$ , then this code is a binary linear  $(n, n - 1)$  code with parity-check matrix  $H = (11 \dots 1)$ .

**Example 9.4 (Repetition code)**

In a repetition code each code word consists of any one message symbol  $a_1$  and  $n - 1$  control symbols  $c_2 = \dots = c_n$  all equal to  $a_1$ . That is  $a_1$  is repeated  $n - 1$  times.

This is a linear  $(n, 1)$  code with parity-check matrix  $H = (-1, I_{n-1})$

The parity-check equations  $H\vec{e}^T = \vec{0}$  with  $H = (A, I_{n-k})$  imply:

$$\begin{aligned}\vec{e}^T &= [I_k - A]\vec{a}^T \\ &= [\vec{a}(I_k - A^T)]^T\end{aligned}$$

where  $\vec{a} = a_1, \dots, a_k$  is the message and  $\vec{c} = c_1, \dots, c_n$  is the codeword. This leads to the following definition:

**Definition**

The  $k \times n$  matrix  $G = (I_k, -A^T)$  is called canonical generator matrix of a linear  $(n, k)$  code with parity-check matrix  $H = (A, I_{n-k})$ .

From  $H\vec{e}^T = \vec{0}$  and  $\vec{c} = \vec{a}G$  it follows that  $H$  and  $G$  are related by:

$$GH^T = \vec{0}$$

The code  $C$  is equal to the row space of the canonical generator matrix  $G$ . More generally, any  $k \times n$  matrix  $G$  whose row space is equal to  $C$  is called a generator matrix of  $C$ .

**Example Generator Matrix**

Let  $H$  be the following  $3 \times 7$  matrix over  $\mathbb{F}_2$ :

$$H = [1 \ 0 \ 1 \ 1 \ 1 \ 0 \ 0 \ 1 \ 1 \ 0 \ 1 \ 0 \ 1 \ 0 \ 1 \ 1 \ 0 \ 0 \ 0 \ 1]$$

The canonical generator matrix for the code defined by  $H$  is given by:

$$G = [1 \ 0 \ 0 \ 0 \ 1 \ 1 \ 1 \ 0 \ 1 \ 0 \ 0 \ 0 \ 1 \ 1 \ 0 \ 0 \ 1 \ 0 \ 0 \ 0 \ 1 \ 1 \ 1 \ 0]$$

**Definition (error word/vector)**

If  $\vec{c}$  is a code word and  $\vec{y}$  is the received word after communication through a "noisy" channel, then  $\vec{e} = \vec{y} - \vec{c} = e_1 \dots e_n$  is called the **error word** or the **error vector**

**Definition**

Let  $\vec{x}, \vec{y}$  be two vectors in  $\mathbb{F}_q^n$ . Then:

- The *Hamming distance*  $d(\vec{x}, \vec{y})$  between  $\vec{x}$  and  $\vec{y}$  is the number of coordinates in which  $\vec{x}$  and  $\vec{y}$  differ:
- The *Hamming weight*  $w(\vec{x})$  of  $\vec{x}$  is the number of nonzero coordinates of  $\vec{x}$ .

Thus  $d(\vec{x}, \vec{y})$  gives the number of errors if  $\vec{x}$  is the transmitted code word and  $\vec{y}$  is the received word. It follows immediately that  $w(\vec{x}) = d(\vec{x}, \vec{0})$  and  $d(\vec{x}, \vec{y}) = w(\vec{x} - \vec{y})$ .

**Lemma Hamming distance**

The Hamming distance is a metric on  $\mathbb{F}_q^n$ . That is for all  $\vec{x}, \vec{y}, \vec{z} \in \mathbb{F}_q^n$  we have:

- $d(\vec{x}, \vec{y}) = 0$  if and only if  $\vec{x} = \vec{y}$
- $d(\vec{x}, \vec{y}) = d(\vec{y}, \vec{x})$
- $d(\vec{x}, \vec{z}) = d(\vec{x}, \vec{y}) + d(\vec{y}, \vec{z})$

In decoding received words  $\vec{y}$ , one usually tries to find the code word  $\vec{c}$  such that  $w(\vec{y} - \vec{c})$  is as small as possible, that is, one assumes that it is more likely that few errors have occurred rather than many. Thus in the decoding we are looking for a code word  $\vec{c}$  that is closest to  $\vec{y}$  according to the Hamming distance.

This rule is called *nearest neighbor decoding*.

**Definition (t-error-correction)**

For  $t \in \mathbb{N}$  a code  $C \subseteq \mathbb{F}_q^n$  is called a *t-error-correcting* if for any  $\vec{y} \in \mathbb{F}_q^n$  there is at most one  $\vec{c} \in C$  such that  $d(\vec{y}, \vec{c}) \leq t$

If  $\vec{c} \in C$  is transmitted and at most  $t$  errors occur, then we have  $d(\vec{y}, \vec{c}) \leq t$  for the received word  $\vec{y}$ . If  $C$  is *t-error-correcting* then for all other code words  $\vec{z} \neq \vec{c}$  we have  $d(\vec{y}, \vec{z}) > t$  which means that  $\vec{c}$  is closest to  $\vec{y}$  and **nearest neighbor decoding** gives the correct result. Therefore, one aim in coding theory is to construct codes with code words "far apart". On the other hand, one tries to transmit as much information as possible. To reconcile these two aims is one of the problems of coding.

**Def: Minimum distance of the linear code.**

The number:

$$d_C = \min_{\vec{u}, \vec{v} \in C, \vec{u} \neq \vec{v}} d(\vec{u}, \vec{v}) = \min_{\vec{0} \neq \vec{c} \in C} w(\vec{c})$$

is called the minimum distance of the linear code  $C$ .

**Theorem**

A code  $C$  with minimum distance  $d_C$  can correct up to  $t$  errors if  $d_C \geq 2t + 1$

**Proof:**

A ball  $B_t(x)$  of radius  $t$  and center  $x \in \mathbb{F}_q^n$  consists of all vectors  $\vec{y} \in \mathbb{F}_q^n$  such that  $d(\vec{x}, \vec{y}) \leq t$ . The nearest neighbor decoding rule ensures that each received word with  $t$  or fewer errors must be in a ball of radius  $t$  and center the transmitted code word. To correct  $t$  errors, the balls with code words  $\vec{x}$  as centers must not overlap. If  $\vec{u} \in B_t(\vec{x})$  and  $\vec{u} \in B_t(\vec{y})$ ,  $\vec{x}, \vec{y} \in C$ ,  $\vec{x} \neq \vec{y}$  then:

$$d(\vec{x}, \vec{y}) \leq d(\vec{x}, \vec{u}) + d(\vec{u}, \vec{y}) \leq 2t$$

a contradiction to  $d_C \geq 2t + 1$

**Example**

Let  $H$  be the following  $3 \times 7$  matrix over  $\mathbb{F}_2$ :

$$H = [1 \ 0 \ 1 \ 1 \ 1 \ 0 \ 0 \ 1 \ 1 \ 0 \ 1 \ 0 \ 1 \ 0 \ 1 \ 1 \ 0 \ 0 \ 0 \ 1]$$

The code has minimum distance  $d_C = 3$  and therefore can correct one error.

**Lemma**

This lemma is useful in determining the minimum distance of a code.

A linear code  $C$  with parity-check matrix  $H$  has minimum distance  $d_C \leq s + 1$  if and only if any  $s$  columns of  $H$  are linearly independent.

**Proof:**

Assume there are  $s$  linearly dependent columns of  $H$ , then  $H\vec{c}^T = \vec{0}$  and  $w(\vec{c}) \leq s$  for suitable  $\vec{c} \in C$ ,  $\vec{c} \neq \vec{0}$ , hence  $d_C \leq s$ . Similarly, if any  $s$  columns of  $H$  are linearly independent, then there is no  $\vec{c} \in C$ ,  $\vec{c} \neq \vec{0}$  of weight  $\leq s$  hence  $d_C \geq s + 1$ .

Next we describe a simple decoding algorithm for linear codes. Let  $C$  be a linear  $(n, k)$  code over  $\mathbb{F}_q$ . The vector space  $\mathbb{F}_q^n / C$  consists of all cosets  $\vec{a} + C = \vec{a} + \vec{c} : \vec{c} \in C$  with  $\vec{a} \in \mathbb{F}_q^n$

Each coset contains  $q^k$  vectors and  $\mathbb{F}_q^n$  can be regarded as being partitioned into cosets of  $C$ :

$$\mathbb{F}_q^n = (\vec{a}^{(0)} + C) \cup (\vec{a}^{(1)} + C) \cup \dots \cup (\vec{a}^{(s)} + C)$$

where  $\vec{a}^{(0)} = \vec{0}$  and  $s = q^{n-k} - 1$ . A received vector  $\vec{y}$  must be in one of the cosets, say in  $\vec{a}^{(i)} + C$ .

If the code word  $\vec{c}$  was transmitted, then the error is given by  $\vec{e} = \vec{y} - \vec{c} = \vec{a}^{(i)} + \vec{z} \in \vec{a}^{(i)} + C$  for a suitable  $\vec{z} \in C$ . This leads to the following decoding scheme.

**Decoding of Linear Codes.**

All possible error vectors  $\vec{e}$  of a received vector  $\vec{y}$  are the vectors in the coset of  $\vec{y}$ . The most likely error vector is the vector  $\vec{e}$  with minimum weight in the coset of  $\vec{y}$ . Thus we decode  $\vec{y}$  as  $\vec{x} = \vec{y} - \vec{e}$ . The implementation of this procedure can be facilitated by the *coset-leader algorithm* for error correction of linear codes.

**Binary Hamming Code**

Definition: A binary  $C_m$  of length  $n = 2^m - 1$ ,  $m \geq 2$  with and  $m \times (2^m - 1)$  parity-check matrix  $H$  is called a binary Hamming Code if the columns of  $H$  are binary representations of the integers  $1, 2, \dots, 2^m - 1$ .

**Lemma**

$C_m$  is a 1-error-correcting code of dimension  $2^m - m - 1$

**Proof**

By definition of the parity check matrix  $H$  of  $C_m$ , the rank of  $H$  is  $m$ . Also any two columns of  $H$  are linearly independent. Since  $H$  contains with any two of its columns also their sum, the minimum distance of  $C_m$  equals 3. Thus  $C_m$  is 1-error-correcting.

**Example:**

Let  $C_3$  be the  $(7, 4)$  Hamming Code with parity-check matrix:

$$H = [0 \ 0 \ 0 \ 1 \ 1 \ 1 \ 1 \ 0 \ 1 \ 1 \ 0 \ 0 \ 1 \ 1 \ 1 \ 0 \ 1 \ 0 \ 1 \ 0 \ 1]$$

If the syndrome of a received word  $\vec{y}$  is  $S(\vec{y}) = (101)^T$ , then we know that the error must have occurred in the fifth position, since 101 is the binary representation of 5.

**Cyclic codes**

Cyclic codes are special class of linear codes that can be implemented fairly simply and whose mathematical structure is reasonably well known.

**Definition**

A linear  $(n, k)$  code  $C$  over  $\mathbb{F}_q$  is called *cyclic* if  $(a_0, a_1, \dots, a_{n-1}) \in C$  implies  $(a_{n-1}, a_0, \dots, a_{n-2}) \in C$

From now on we impose the restriction  $\gcd(n, q) = 1$  and let  $(x^n - 1)$  be the ideal generated by  $x^n - 1 \in \mathbb{F}_q[x]$ . Then all elements of  $\mathbb{F}_q[x]/(x^n - 1)$  can be represented by polynomials of degree less than  $n$  and clearly this residue class ring is isomorphic to  $\mathbb{F}_q^n$  as a vector space over  $\mathbb{F}_q$ . An isomorphism is given by:

